# ST. ANNE'S COLLEGE OF ENGINEERING AND TECHNOLOGY 

ANGUCHETTYPALAYAM, PANRUTI-607 110 DEPARTMENT OF ELECTRONICS \&

COMMUNICATION ENGINEERING

## Lab Manual



# EC8462 - LINEAR INTEGRATED CIRCUITS LABORATORY 

## EC8462 LINEAR INTEGRATED CIRCUITS LABORATORY <br> LIST OF EXPERIMENTS:

## DESIGN AND TESTING OF

1. Inverting, Non inverting and Differential amplifiers.
2. Integrator and Differentiator.
3. Instrumentation amplifier
4. Active low-pass, High-pass and band-pass filters.
5. Astable \& Monostable multivibrators and Schmitt Trigger using op-amp.
6. Phase shift and Wien bridge oscillators using op-amp.
7. Astable and monostable multivibrators using NE555 Timer.
8. PLL characteristics and its use as Frequency Multiplier.
9. R-2R Ladder Type D - A Converter using Op amp.
10. DC power supply using LM317 and LM723.
11. Study of SMPS.

## SIMULATION USING SPICE

1. Active low-pass, High-pass and band-pass filters using Opamp
2. Astable and Monostable multivibrators using NE555 Timer
3. A/D converters
4. Analog multiplier

## 1. INVERTING, NON-INVERTING AND DIFFERENTIAL AMPLIFIERS.

## 1(a) INVERTING AMPLIFIER

## AIM:

To design an Inverting Amplifier for the given specifications using Op-Amp IC 741.

## APPARATUS REQUIRED:

| S.No | Name of the Apparatus | Range | Quantity |
| :---: | :--- | :--- | :---: |
| 1. | Function Generator | 3 MHz | 1 |
| 2. | CRO | 30 MHz | 1 |
| 3. | Dual RPS | $0-30 \mathrm{~V}$ | 1 |
| 4. | Op-Amp | IC 741 | 1 |
| 5. | Bread Board |  | 1 |
| 6. | Resistors | As required |  |
| 7. | Connecting wires and probes | As required |  |

## THEORY:

The input signal $V_{i}$ is applied to the inverting input terminal through $R_{1}$ and the noninverting input terminal of the op-amp is grounded. The output voltage $V_{o}$ is fed back to the inverting input terminal through the $\mathrm{R}_{\mathrm{f}}-\mathrm{R}_{1}$ network, where $\mathrm{R}_{\mathrm{f}}$ is the feedback resistor. The output voltage is given as,

$$
\mathrm{V}_{\mathrm{o}}=-\mathrm{A}_{\mathrm{CL}} \mathrm{~V}_{\mathrm{i}}
$$

Here the negative sign indicates that the output voltage is $180^{\circ}$ out of phase with the input signal.

## PROCEDURE:

1. Connections are given as per the circuit diagram.
2. $+\mathrm{V}_{\mathrm{cc}}$ and $-\mathrm{V}_{\mathrm{cc}}$ supply is given to the power supply terminal of the Op-Amp IC.
3. By adjusting the amplitude and frequency knobs of the function generator, appropriate input voltage is applied to the inverting input terminal of the Op-Amp.
4. The output voltage is obtained in the CRO and the input and output voltage waveforms are plotted in a graph sheet.

## PIN DIAGRAM:



## CIRCUIT DIAGRAM OF INVERTING AMPLIFIER:



## DESIGN:

Gain of an inverting amplifier $\mathrm{Av}=\mathrm{V}_{\mathrm{o}} / \mathrm{V}_{\text {in }}=-\mathrm{R}_{\mathrm{f}} / \mathrm{R}_{\mathrm{i}}$
The required gain $=10$,
That is $\mathrm{Av}=-\mathrm{Rf} / \mathrm{Ri}_{\mathrm{i}}=10$
Let $\mathrm{R}_{\mathrm{i}}=1 \mathrm{~K} \Omega, \quad$ Then $\mathrm{R}_{\mathrm{f}}=10 \mathrm{~K} \Omega$

## Observations:

Vin $=1 \mathrm{Vpp}$
$\mathrm{Vo}=$ ?
Gain, $\mathrm{Av}=\mathrm{Vo} / \mathrm{Vin}=$ ?
Observed phase difference between the input and the output on the $\mathrm{CRO}=$ ?

## Calculation:

## OBSERVATIONS:

| S.No |  | Input | Output |
| :---: | :--- | :--- | :--- |
| 1. | Amplitude <br> ( No. of div x Volts per <br> div ) |  |  |
| 2. | Time period <br> ( No. of div x Time per <br> div ) |  |  |

## MODEL GRAPH:



## RESULT:

The design and testing of the inverting amplifier is done and the input and output waveforms were drawn.

## VIVA QUESTIONS:

1. What is an op-amp?
2. What is an inverting amplifier?
3. What is the difference between inverting and non inverting amplifier?
4. Define CMRR.
5. Write the equation for gain of an inverting amplifier.

## 1(b) NON - INVERTING AMPLIFIER

## AIM:

To design a Non-Inverting Amplifier for the given specifications using Op-Amp IC 741.

## APPARATUS REQUIRED:

| S.No | Name of the Apparatus | Range | Quantity |
| :---: | :--- | :--- | :---: |
| 1. | Function Generator | 3 MHz | 1 |
| 2. | CRO | 30 MHz | 1 |
| 3. | Dual RPS | $0-30 \mathrm{~V}$ | 1 |
| 4. | Op-Amp | IC 741 | 1 |
| 5. | Bread Board |  | 1 |
| 6. | Resistors | As required |  |
| 7. | Connecting wires and probes | As required |  |

## THEORY:

The input signal $\mathrm{V}_{\mathrm{i}}$ is applied to the non - inverting input terminal of the op-amp. This circuit amplifies the signal without inverting the input signal. It is also called negative feedback system since the output is feedback to the inverting input terminals. The differential voltage $\mathrm{V}_{\mathrm{d}}$ at the inverting input terminal of the op-amp is zero ideally and the output voltage is given as,

$$
\mathrm{V}_{\mathrm{o}}=\mathrm{A}_{\mathrm{CL}} \mathrm{~V}_{\mathrm{i}}
$$

Here the output voltage is in phase with the input signal.

## PROCEDURE:

1. Connections are given as per the circuit diagram.
2. $+\mathrm{V}_{\mathrm{cc}}$ and $-\mathrm{V}_{\mathrm{cc}}$ supply is given to the power supply terminal of the Op-Amp IC.
3. By adjusting the amplitude and frequency knobs of the function generator, appropriate input voltage is applied to the non - inverting input terminal of the Op-Amp.
4. The output voltage is obtained in the CRO and the input and output voltage waveforms are plotted in a graph sheet.

## PIN DIAGRAM:



## CIRCUIT DIAGRAM OF NON INVERITNG AMPLIFIER:

## DESIGN:



Gain of an inverting amplifier $\mathrm{Av}=\mathrm{V}_{\mathrm{o}} / \mathrm{V}_{\mathrm{in}}=1+\mathrm{R}_{\mathrm{f}} / \mathrm{R}_{\mathrm{i}}$,
Let the required gain be 11 ,
Therefore

$$
\begin{aligned}
& \mathrm{Av}=1+\mathrm{R}_{\mathrm{f}} / \mathrm{Ri}_{\mathrm{i}}=11 \\
& \mathrm{R}_{\mathrm{f}} / \mathrm{R}_{\mathrm{i}}=10
\end{aligned}
$$

Take $\mathrm{R}_{\mathrm{i}}=1 \mathrm{~K} \Omega, \quad$ Then $\mathrm{R}_{\mathrm{f}}=10 \mathrm{~K} \Omega$
$\mathrm{Vin}=1 \mathrm{Vpp}$
$\mathrm{Vo}=$ ?
Gain $\mathrm{Av}=\mathrm{Vo} / \mathrm{Vin}=$ ?
Observed phase difference between the input and the output on the $\mathrm{CRO}=$ ?

## CALCULATION:

## OBSERVATIONS:

| S.No |  | Input | Output |
| :---: | :--- | :---: | :---: |
| 1. | Amplitude <br> ( No. of div x Volts per div ) |  |  |
| 2. | Time period <br> (No. of div x Time per div ) |  |  |

## MODEL GRAPH:



## RESULT:

The design and testing of the Non-inverting amplifier is done and the input and output waveforms were drawn.

## VIVA QUESTIONS:

1. What is an op-amp?
2. What is an non- inverting amplifier?
3. What is the difference between inverting and non inverting amplifier?
4. Define CMRR.
5. Write the equation for gain of an non- inverting amplifier.

AIM:
To design a differential Amplifier for the given specifications using Op-Amp IC 741.

## APPARATUS REQUIRED:

| S.No | Name of the Apparatus | Range | Quantity |
| :---: | :--- | :--- | :---: |
| 1. | Function Generator | 3 MHz | 1 |
| 2. | CRO | 30 MHz | 1 |
| 3. | Dual RPS | $0-30 \mathrm{~V}$ | 1 |
| 4. | Op-Amp | IC 741 | 1 |
| 5. | Bread Board |  | 1 |
| 6. | Resistors | As required |  |
| 7. | Connecting wires and probes | As required |  |

## THEORY:

Differential $Z_{\text {in }}$ (between the two input pins) $=R_{1}+R_{2}$ (Note: this is approximate)
For common-mode rejection, anything done to one input must be done to the other. The addition of a compensation capacitor in parallel with Rf, for instance, must be balanced by an equivalent capacitor in parallel with Rg.

Whenever $R_{1}=R_{2 \text { and }} R_{\mathrm{f}}=R_{\mathrm{g}}$, the differential gain is

$$
V_{\text {out }}=A\left(V_{2}-V_{1}\right)_{\text {and }} A \triangleq \frac{R_{\mathrm{f}}}{R_{1}}
$$

When $R_{1}=R_{\text {fand }} R_{2}=R_{\text {gthe differential gain is } \mathrm{A}=1 \text { and the circuit acts as a differential }}$ follower:

$$
V_{\mathrm{out}}=V_{2}-V_{1}
$$

## PROCEDURE:

1. Connections are given as per the circuit diagram.
2. $+\mathrm{V}_{\mathrm{cc}}$ and $-\mathrm{V}_{\mathrm{cc}}$ supply is given to the power supply terminal of the Op-Amp IC.
3. By adjusting the amplitude and frequency knobs of the function generator, appropriate input voltage is applied to the input terminal of the Op-Amp.
4. The output voltage is obtained in the CRO and the input and output voltage waveforms are plotted in a graph sheet.


## CIRCUIT DIAGRAM OF DIFFERENTIAL AMPLIFIER:



## DESIGN:

$$
V_{\text {out }}=\frac{\left(R_{\mathrm{f}}+R_{1}\right) R_{\mathrm{g}}}{\left(R_{\mathrm{g}}+R_{2}\right) R_{1}} V_{2}-\frac{R_{\mathrm{f}}}{R_{1}} V_{1}
$$

Differential $Z_{\text {in }}$ (between the two input pins) $=R_{1}+R_{2}$ (Note: this is approximate)
For common-mode rejection, anything done to one input must be done to the other. The addition of a compensation capacitor in parallel with Rf, for instance, must be balanced by an equivalent capacitor in parallel with Rg.

Whenever $R_{1}=R_{2 \text { and }} R_{\mathrm{f}}=R_{\mathrm{g} \text {, the differential gain is }}$

$$
V_{\mathrm{out}}=A\left(V_{2}-V_{1}\right)_{\text {and }} A \triangleq \frac{R_{\mathrm{f}}}{R_{1}}
$$

When $R_{1}=R_{\text {fand }} R_{2}=R_{\text {gthe differential gain is } \mathrm{A}=1}$ and the circuit acts as a differential follower:

$$
V_{\mathrm{out}}=V_{2}-V_{1}
$$

## OBSERVATIONS:

| S.No |  | Input | Output |
| :---: | :--- | :---: | :---: |
| 1. | Amplitude in differential mode <br> (No. of div x Volts per div ) |  |  |
| 2 | Amplitude in common mode <br> (No. of div x Volts per div ) |  |  |
| 3 | Time period <br> (No. of div x Time per div ) |  |  |

## MODEL GRAPH:



## RESULT:

The design and testing of the differential amplifier is done and the input and output waveforms were drawn.

## 2(a) INTEGRATOR:

## AIM:

To design an Integrator circuit for the given specifications using Op-Amp IC 741.

## APPARATUS REQUIRED:

| S.No | Name of the Apparatus | Range | Quantity |
| :---: | :--- | :--- | :---: |
| 1. | Function Generator | 3 MHz | 1 |
| 2. | CRO | 30 MHz | 1 |
| 3. | Dual RPS | $0-30 \mathrm{~V}$ | 1 |
| 4. | Op-Amp | IC 741 | 1 |
| 5. | Bread Board |  | 1 |
| 6. | Resistors |  |  |
| 7. | Capacitors |  |  |
| 8. | Connecting wires and probes | As required |  |

## THEORY:

A circuit in which the output voltage waveform is the integral of the input voltage waveform is the integrator. Such a circuit is obtained by using a basic inverting amplifier configuration if the feedback resistor $\mathrm{R}_{\mathrm{f}}$ is replaced by a capacitor $\mathrm{C}_{\mathrm{f}}$. The expression for the output voltage is given as,

$$
V_{o}=-\left(1 / R_{f} C_{1}\right) \int V_{i} d t
$$

Here the negative sign indicates that the output voltage is $180^{\circ}$ out of phase with the input signal. Normally between $f_{a}$ and $f_{b}$ the circuit acts as an integrator. Generally, the value of $f_{a}<$ $f_{b}$. The input signal will be integrated properly if the Time period $T$ of the signal is larger than or equal to $R_{f} C_{f}$. That is, $\quad T \geq R_{f} C_{f}$

The integrator is most commonly used in analog computers and ADC and signal-wave shaping circuits.

## PIN DIAGRAM:



## CIRCUIT DIAGRAM OF INTEGRATOR:



## DESIGN:

Given $\mathrm{f}=1 \mathrm{KHz}$
So $T=1 / \mathrm{f}=1 \mathrm{~ms}$
Design equation is $\mathrm{T}=2 \pi \mathrm{R}_{\mathrm{i}} \mathrm{C}$
Let $\mathrm{C}=0.01 \mu \mathrm{~F}$
Then $\mathrm{R}_{\mathrm{i}}=15 \mathrm{~K} \Omega$
Take $\mathrm{R}_{\mathrm{f}}=10 \mathrm{R}_{\mathrm{i}}=150 \mathrm{~K} \Omega$
[ To obtain the output of an Integrator circuit with component values $\mathrm{R}_{1} \mathrm{C}_{\mathrm{f}}=0.1 \mathrm{~ms}, \mathrm{R}_{\mathrm{f}}=10 \mathrm{R}_{1}$ and $\mathrm{C}_{\mathrm{f}}=0.01 \mu \mathrm{~F}$ and also if 1 V peak square wave at 1000 Hz is applied as input.]

We know the frequency at which the gain is $0 d B, f_{b}=1 /\left(2 \pi R_{1} C_{f}\right)$
Therefore $\mathrm{f}_{\mathrm{b}}=$ $\qquad$
Since $f_{b}=10 f_{a}$, and also the gain limiting frequency $f_{a}=1 /\left(2 \pi R_{f} C_{f}\right)$
We get, $\mathrm{R}_{1}=$ $\qquad$ and hence $R_{f}=$ $\qquad$

## PROCEDURE:

1. Connections are given as per the circuit diagram.

2 . $+\mathrm{V}_{\mathrm{cc}}$ and $-\mathrm{V}_{\mathrm{cc}}$ supply is given to the power supply terminal of the Op-Amp IC.
3. By adjusting the amplitude and frequency knobs of the function generator, appropriate input voltage is applied to the inverting input terminal of the Op-Amp.
4. The output voltage is obtained in the CRO and the input and output voltage waveforms are plotted in a graph sheet.

## OBSERVATIONS:

| S.No |  | amplitude | Time period |
| :---: | :--- | :---: | :---: |
| 1. | Sine wave input |  |  |
| Cosine wave output |  |  |  |$\quad$|  |
| :--- |
| 2. | | Square wave input |
| :--- |
| Spike wave output |$\quad$|  |
| :--- | :--- |

## MODEL GRAPH:




## CALCULATION:

## RESULT:

The design of the Integrator circuit was done and the input and output waveforms were obtained.

## 2(b) DIFFERENTIATOR

AIM:
To design a Differentiator circuit for the given specifications using Op-Amp IC 741.

## APPARATUS REQUIRED:

| S.No | Name of the Apparatus | Range | Quantity |
| :---: | :--- | :--- | :---: |
| 1. | Function Generator | 3 MHz | 1 |
| 2. | CRO | 30 MHz | 1 |
| 3. | Dual RPS | $0-30 \mathrm{~V}$ | 1 |
| 4. | Op-Amp | IC 741 | 1 |
| 5. | Bread Board |  | 1 |
| 6. | Resistors |  |  |
| 7. | Capacitors | As required |  |
| 8. | Connecting wires and probes |  |  |

## THEORY:

The differentiator circuit performs the mathematical operation of differentiation; that is, the output waveform is the derivative of the input waveform. The differentiator may be constructed from a basic inverting amplifier if an input resistor $\mathrm{R}_{1}$ is replaced by a capacitor $\mathrm{C}_{1}$. The expression for the output voltage is given as,

$$
V_{o}=-R_{f} C_{1}\left(d V_{i} / d t\right)
$$

Here the negative sign indicates that the output voltage is $180^{0}$ out of phase with the input signal. A resistor $\mathrm{R}_{\text {comp }}=\mathrm{R}_{\mathrm{f}}$ is normally connected to the non-inverting input terminal of the opamp to compensate for the input bias current. A workable differentiator can be designed by implementing the following steps:

1. Select $f_{a}$ equal to the highest frequency of the input signal to be differentiated. Then, assuming a value of $\mathrm{C}_{1}<1 \mu \mathrm{~F}$, calculate the value of $\mathrm{R}_{\mathrm{f}}$.
2. Choose $f_{b}=20 f_{a}$ and calculate the values of $R_{1}$ and $C_{f}$ so that $R_{1} C_{1}=R_{f} C_{f}$.

The differentiator is most commonly used in wave shaping circuits to detect high frequency components in an input signal and also as a rate-of-change detector in FM modulators.

## PIN DIAGRAM:



## CIRCUIT DIAGRAM OF DIFFERENTIATOR:



## DESIGN :

Given $\mathrm{f}=1 \mathrm{KHz}$
So $\mathrm{T}=1 / \mathrm{f}=1 \mathrm{~ms}$
Design equation is $\mathrm{T}=2 \pi \mathrm{R}_{\mathrm{f}} \mathrm{C}$
Let $\mathrm{C}=0.01 \mu \mathrm{~F}$
Then $\mathrm{R}_{\mathrm{f}}=15 \mathrm{~K} \Omega$
Let $\mathrm{Ri}=\mathrm{Rf} / 10=1.5 \mathrm{~K} \Omega$
[ To design a differentiator circuit to differentiate an input signal that varies in frequency from 10 Hz to about 1 KHz . If a sine wave of 1 V peak at 1000 Hz is applied to the differentiator, draw its output waveform.]
Given $f_{a}=1 \mathrm{KHz}$
We know the frequency at which the gain is $0 \mathrm{~dB}, \mathrm{fa}=1 /\left(2 \pi \mathrm{R}_{\mathrm{f}} \mathrm{C}_{1}\right)$
Let us assume $\mathrm{C}_{1}=0.1 \mu \mathrm{~F}$; then
$\mathrm{R}_{\mathrm{f}}=$ $\qquad$
Since $f_{b}=20 f_{a}, f_{b}=20 \mathrm{KHz}$
We know that the gain limiting frequency $f_{b}=1 /\left(2 \pi R_{1} C_{1}\right)$
Hence $\mathrm{R}_{1}=$ $\qquad$
Also since $\mathrm{R}_{1} \mathrm{C}_{1}=\mathrm{R}_{\mathrm{f}} \mathrm{C}_{\mathrm{f}} ; \mathrm{Cf}=$ $\qquad$
Given $\mathrm{V}_{\mathrm{p}}=1 \mathrm{~V}$ and $\mathrm{f}=1000 \mathrm{~Hz}$, the input voltage is $\mathrm{V}_{\mathrm{i}}=\mathrm{V}_{\mathrm{p}} \sin \omega \mathrm{t}$
We know $\omega=2 \pi f$
Hence

$$
\begin{aligned}
\mathrm{Vo}=-\mathrm{R}_{\mathrm{f}} & \mathrm{C}_{1}\left(\mathrm{dV}_{\mathrm{i}} / \mathrm{dt}\right) \\
& =-0.94 \cos \omega \mathrm{t}
\end{aligned}
$$

## PROCEDURE:

1. Connections are given as per the circuit diagram.
2. $+\mathrm{V}_{\mathrm{cc}}$ and $-\mathrm{V}_{\mathrm{cc}}$ supply is given to the power supply terminal of the Op-Amp IC.
3. By adjusting the amplitude and frequency knobs of the function generator, appropriate input voltage is applied to the inverting input terminal of the Op-Amp.
4. The output voltage is obtained in the CRO and the input and output voltage waveforms are plotted in a graph sheet.

## OBSERVATIONS:

| S.No |  | amplitude | Time period |
| :---: | :--- | :---: | :---: |
| 1. | Sine wave input |  |  |
| Cosine wave output |  |  |  |$\quad$|  |
| :--- |
| 2. |
| Square wave input |
| Spike wave output |

## MODEL GRAPH:




## CALCULATION:

## RESULT:

The design of the Differentiator circuit was done and the input and output waveforms were obtained.
VIVA QUESTIONS.

1. Define an integrator.
2. State the applications of an integrator.
3. What is a differentiator?
4. What are the steps to design a differentiator?
5. What are the steps to design an integrator?

## 3. INSTRUMENTATION AMPLIFIER.

AIM:

To design a instrumentation Amplifier for the given specifications using Op-Amp IC 741.

## APPARATUS REQUIRED:

| S.No | Name of the Apparatus | Range | Quantity |
| :---: | :--- | :--- | :---: |
| 1. | Function Generator | 3 MHz | 1 |
| 2. | CRO | 30 MHz | 1 |
| 3. | Dual RPS | $0-30 \mathrm{~V}$ | 1 |
| 4. | Op-Amp | IC 741 | 1 |
| 5. | Bread Board |  | 1 |
| 6. | Resistors | As required |  |
| 7. | Connecting wires and probes | As required |  |

## THEORY:

An instrumentation (or instrumentational) amplifier is a type of differential amplifier that has been outfitted with input buffers, which eliminate the need for input impedance matching and thus make the amplifier particularly suitable for use in measurement and test equipment. Additional characteristics include very low DC offset, low drift, low noise, very high open-loop gain, very high common-mode rejection ratio, and very high input impedances. Instrumentation amplifiers are used where great accuracy and stability of the circuit both short- and long-term are required.

The most commonly used instrumentation amplifier circuit is shown in the figure. The gain of the circuit is

$$
\frac{V_{\text {out }}}{V_{2}-V_{1}}=\left(1+\frac{2 R_{1}}{R_{\text {gain }}}\right) \frac{R_{3}}{R_{2}}
$$

The rightmost amplifier, along with the resistors labelled $R_{2}$ and $R_{3}$ is just the standard differential amplifier circuit, with gain $=R_{3} / R_{2}$ and differential input resistance $=2 \cdot R_{2}$. The two amplifiers on the left are the buffers. With $R_{\text {gain }}$ removed (open circuited), they are simple unity gain buffers; the circuit will work in that state, with gain simply equal to $R_{3} / R_{2}$ and high input impedance because of the buffers. The buffer gain could be increased by putting resistors between the buffer inverting inputs and ground to shunt away some of the negative feedback; however, the single resistor $R_{\text {gain }}$ between the two inverting inputs is a much more elegant method: it increases the differential-mode gain of the buffer pair while leaving the common-mode gain equal to 1 .

## PROCEDURE:

1. Connections are given as per the circuit diagram.
2. $+\mathrm{V}_{\mathrm{cc}}$ and $-\mathrm{V}_{\mathrm{cc}}$ supply is given to the power supply terminal of the Op-Amp IC.
3. By adjusting the amplitude and frequency knobs of the function generator, appropriate input voltage is applied to the input terminal of the Op-Amp.
4. The output voltage is obtained in the CRO and the input and output voltage waveforms are plotted in a graph sheet.

## PIN DIAGRAM:



## CIRCUIT DIAGRAM OF INSTRUMENTATION AMPLIFIER:



## OBSERVATIONS:

| S.No |  | Amplitude | Time period |
| :---: | :--- | :--- | :--- |
| 1. | Input |  |  |
| 2 | Output |  |  |

## RESULT:

The design and testing of the instrumentation amplifier is done and the input and output waveforms were drawn.

## VIVA QUESTIONS:

1. What are the features of instrumentation amplifier?
2. What are the applications of instrumentation amplifier?
3. What is an instrumentation amplifier?
4. Write the expression for output voltage.
5. What is the use of transducer in an instrumentation amplifier?

## 4. ACTIVE LOWPASS, HIGHPASS AND BANDPASS FILTERS

## Aim:-

To design and test the frequency response of a second order LPF, HPF and BPF.

## Components Required:-

| S.No | Components | Range | Quantity |
| :--- | :--- | :--- | :--- |
| 1. | Op-amp | IC 741 | 1 |
| 2. | Resistors |  |  |
| 3. | Capacitor | O.01 $\mu \mathrm{f}, \mathrm{O} .05 \mu \mathrm{f}$ | 2 each |
| 4. | CRO |  | 1 |
| 5. | Power Supply | $\pm 15 \mathrm{~V}$ | 1 |
| 6. | Probe |  | 2 |
| 7. | Bread Board |  | 1 |

## Theory:-

## LPF:-

A LPF allows only low frequency signals up to a certain break-point $f_{H}$ to pass through, while suppressing high frequency components. The range of frequency from 0 to higher cut off frequency $f_{H}$ is called pass band and the range of frequencies beyond $f_{H}$ is called stop band.

The following steps are used for the design of active LPF.

1. The value of high cut off frequency $f_{H}$ is chosen.
2. The value of capacitor C is selected such that its value is $\leq 1 \mu \mathrm{~F}$.
3. By knowing the values of $f_{H}$ and $C$, the value of $R$ can be calculated using

$$
f_{H}=\frac{1}{2 \pi R C}
$$

4. Finally the values of $\mathrm{R}_{1}$ and $\mathrm{R}_{\mathrm{f}}$ are selected depending on the designed pass band gain by using $A=1+\left(\frac{R_{f}}{R_{1}}\right)$

## Circuit Diagram:-

## Second Order LPF:



## Design:-

Second order:-
Given frequency, $\mathrm{f}_{\mathrm{H}}=2 \mathrm{KHz}$ and gain $=2$
Let

$$
\mathrm{C}=0.01 \mu \mathrm{f}
$$

The frequency, $\mathrm{f}_{\mathrm{H}}=\frac{1}{2 \pi \sqrt{\left(2 \times 10^{3}\right)\left(0.01 \times 10^{-6}\right)}}$

$$
\begin{aligned}
& \text { Set, } \quad R_{2}=R_{3}=R \\
& C_{2}=C_{3}=C \\
& \therefore f_{H}=\frac{1}{2 \pi R C}
\end{aligned}
$$

Tabulation
Second order LPF Vin=1V

| S.No | Frequency (Hz) | O/p voltage(v) | Gain=Vo/Vin | Gain=20log(Vo/Vin) |
| :--- | :--- | :--- | :--- | :--- |
|  |  |  |  |  |
|  |  |  |  |  |
|  |  |  |  |  |
|  |  |  |  |  |

## Model graph:-



## Second order HPF:

## Theory:-

The high pass filter is the complement of the low pass filter. Thus the high pass filter can be obtained by interchanging R and C in the circuit of low pass configuration. A high pass filter allows only frequencies above a certain bread point to pass through and at terminates the low frequency components. The range of frequencies beyond its lower cut off frequency $f_{L}$ is called stop band.

## Circuit Diagram:-

## Second Order HPF:



## Design:-

$$
\begin{aligned}
& f_{L}=2 \mathrm{KHZ} \\
& C=0.01 \mu \mathrm{H} \\
& \text { Gain, Av=2 } \\
& f_{L}=\frac{1}{2 \pi \sqrt{R_{2} R_{3} C_{2} C_{3}}} \\
& L_{e t} R_{2}=R_{3}=R \\
& R_{2}=R_{3}=C \\
& R_{2}=R_{3}=7.95 k \Omega \\
& A=1+\frac{1}{2 \pi f Z} \\
& R_{1}=2 \\
& \therefore R_{f}=R_{1}=10 k \Omega(\text { given }
\end{aligned}
$$

## Observation:

| S.No | Frequency (Hz) | O/p voltage(v) | Gain=Vo/Vin | Gain=20log(Vo/Vin) |
| :--- | :--- | :--- | :--- | :--- |
|  |  |  |  |  |
|  |  |  |  |  |
|  |  |  |  |  |
|  |  |  |  |  |

## Model graph:-



## Procedure:-

LPF:-

1. Connections are given as per the circuit diagram.
2. Input signal is connected to the circuit from the signal generator.
3. The input and output signals of the filter channels 1 and 2 of the CRO are connected.
4. Suitable voltage sensitivity and time-base on CRO is selected.
5. The correct polarity is checked.
6. The above steps are repeated for second order filter.

## HPF

1. Connections are given as per the circuit diagram.
2. Input signal is connected to the circuit from the signal generator.
3. The input and output signals of the filter channels 1 and 2 of the CRO are connected.
4. Suitable voltage sensitivity and time-base on CRO is selected.
5. The correct polarity is checked.
6. The above steps are repeated for second order filter.

## Theory:-

## BPF:-

The BPF is the combination of high and low pass filters and this allows a specified range of frequencies to pass through. It has two stop bands in range of frequencies between 0 to $f_{\mathrm{L}}$ and beyond $f_{H}$. The band $b / w f_{L}$ and $f_{H}$ is called pass band. Hence its bandwidth is $\left(f_{L}-f_{H}\right)$. This filter has a maximum gain at the resonant frequency ( $\mathrm{f}_{\mathrm{r}}$ ) which is defined as

$$
f_{r}=\sqrt{f_{H} f_{L}}
$$

The figure of merit (or) quality factor Q is given by

$$
Q=\frac{f_{r}}{f_{H}-f_{L}}=\frac{f_{r}}{B W}
$$

## Circuit Diagram:-

## BPF



## Model graph:-

BPF:-


## Tabulation:-

BPF Vin=50mv

| S.No | Frequency (Hz) | Vo(volts) | Gain=20log(Vo/Vin) |
| :--- | :--- | :--- | :--- |
|  |  |  |  |
|  |  |  |  |
|  |  |  |  |

## Procedure:

 BPF:-1. The input signal is connected to the circuit from the signal generator.
2. The input and output signals are connected to the filter.
3. The suitable voltage is selected.
4. The correct polarity is checked.
5. The steps are repeated.

## Result:-

Thus the frequency response of second order LPF,HPF and BSF filter was designed and tested.

## 5. ASTABLE AND MONOSTABLE MULTIVIBRATORS AND SCHMITT TRIGGER USING OP-AMP

## Aim:

To design and test an astable and monostable multivibrators and schmitt trigger using opamp

## Apparatus Required:

| S.No | Component | Range | Quantity |
| :---: | :--- | :--- | :---: |
| 1. | Op amp | IC 741 | 1 |
| 2. | DTS | $(0-30) \mathrm{V}$ | 1 |
| 3. | CRO |  | 1 |
| 4. | Resistor | - | 1 |
| 5. | Capacitors | - | - |
| 6. | Diode | - | 2 |
| 7. | Probes |  | 1 |

## Design:

## 1. Monostable Multivibrators:

$\beta=\mathrm{R}_{2} / \mathrm{R}_{1}+\mathrm{R}_{2}\left[\beta=0.5 \& \mathrm{R}_{1}=10 \mathrm{~K}\right]$
Find $\mathrm{R}_{2}=\quad ; \mathrm{R} 3=1 \mathrm{~K} ; \mathrm{R} 4=10 \mathrm{~K}$;
Let $\mathrm{F}=\ldots \quad \mathrm{KHz} ; \mathrm{C}=1 \mathrm{mfd} ; \mathrm{C} 4=0.1 \mathrm{mfd}$
Pulse width, $\mathrm{T}=0.69 \mathrm{RC}$
Find $\mathrm{R}=$
Time Period $\mathrm{T}=0.69 \mathrm{RC}$
Let $\mathrm{T}=1 \mathrm{~ms}$; and $\mathrm{C}=0.1 \mu \mathrm{~F}$.
Then $R=15 \mathrm{~K} \Omega$
Feedback factor $\beta=\mathrm{R}_{2} /\left(\mathrm{R}_{1}+\mathrm{R}_{2}\right)$
Let $\beta=0.5$ or $1 / 2$
$\mathrm{R}_{2} /\left(\mathrm{R}_{1}+\mathrm{R}_{2}\right)=1 / 2$
If $\mathrm{R}_{1}=10 \mathrm{~K} \Omega ; \quad \mathrm{R}_{2}=10 \mathrm{~K} \Omega$

## For triggering circuit

$\mathrm{R}_{\mathrm{d}} \mathrm{C}_{\mathrm{d}}=0.0016 \mathrm{t}$
Let $\mathrm{t}=3 \mathrm{~ms}$ and $\mathrm{C}_{\mathrm{d}}=0.01 \mu \mathrm{~F}$; then $\mathrm{Rd}=470 \Omega$

## Circuit Diagram



Tabulation:

| S.No |  | amplitude | Time period |  |
| :--- | :--- | :--- | :--- | :--- |
| Ton |  |  |  |  |
|  |  |  |  |  |

## Model graph:



## Procedure:

1. Make the connections as shown in circuit diagram.
2. A trigger pulse is given through differentiator circuit through pin no. 3
3. Observe the pulse waveform at pin no. 6 using CRO and note down the time period.
4. Plot the waveform on the graph.

## 2. Astable Multivibrators:

Design: $\quad T=2 R C$

$$
\mathrm{R}_{1}=1.16 \mathrm{R}_{2}
$$

Given $\mathrm{f}_{\mathrm{O}}=$ $\qquad$ KHz

Frequency of Oscillation fo $=1 / 2 \mathrm{RC}$ if $\mathrm{R}_{1}=1.16 \mathrm{R}_{2}$
Let $\mathrm{R}_{2}=10 \mathrm{~K} \Omega$

$$
\mathrm{R}_{1}=10 * 1.16=11.6 \mathrm{~K} \Omega
$$

Let $\mathrm{C}=0.05 \mu \mathrm{~F}$

$$
\mathrm{R}=1 / 2 \mathrm{fC}=1 /\left(2 * 1 * 10^{3} * 0.05 * 10^{-6}\right)=
$$

Circuit Diagram


SYMMETRICAL ASTABLE MULTIVIBRATOR

Tabulation:

| S.No |  |  |  |
| :--- | :--- | :--- | :--- |
|  |  |  |  |
|  |  |  |  |

## Model graph



## Procedure:

1. Make the connections as shown in the circuit diagram
2. Keep the CRO channel switch in ground and adjust the horizontal line on the x axis so that it coincides with the central line.
3. Select the suitable voltage sensitivity and time base on the CRO.
4. Check for the correct polarity of the supply voltage to op-amp and switch on power supply to the circuit.
5. Observe the waveform at the output and across the capacitor. Measure the frequency of oscillation and the amplitude. Compare with the designed value.
6. Plot the Waveform on the graph.

## Design:

$\mathrm{f}=1 \mathrm{KHz}$
$\mathrm{T}=1 / \mathrm{f}=1 \mathrm{~ms}$
$\beta=\mathrm{R}_{1} /\left(\mathrm{R}_{1}+\mathrm{R}_{2}\right)$
Let $\mathrm{R}_{1}=10 \mathrm{~K} \Omega$, and $\mathrm{R}_{2}=10 \mathrm{~K} \Omega$
Then $\beta=0.5$
Therefore $\mathrm{T}=2.2 \mathrm{RC}=1 \mathrm{~ms}$

## Schmitt Trigger:

## Design

$\mathrm{V}_{\mathrm{CC}}=12 \mathrm{~V} ; \mathrm{V}_{\mathrm{SAT}}=0.9 \mathrm{~V}_{\mathrm{CC}} ; \mathrm{R} 1=47 \mathrm{~K} \Omega ; \mathrm{R} 2=120 \Omega$

$$
\mathrm{V}_{\mathrm{UT}}=+\left[\mathrm{V}_{\mathrm{SAT}} \mathrm{R}_{2}\right] /\left[\mathrm{R}_{1}+\mathrm{R}_{2}\right] \& \mathrm{~V}_{\mathrm{LT}}=-\left[\mathrm{V}_{\mathrm{SAT}} \mathrm{R}_{2}\right] /\left[\mathrm{R}_{1}+\mathrm{R}_{2}\right] \& \text { HYSTERSIS }[\mathrm{H}]=\mathrm{V}_{\mathrm{UT}}-\mathrm{V}_{\mathrm{LT}}
$$

## Circuit Diagram



Tabulation:


Model Graph


## Procedure

1. Connect the circuit as shown in the circuit
2. Set the input voltage as 5 V ( $\mathrm{p}-\mathrm{p}$ ) at 1 KHz . (Input should be always less than $\mathrm{V}_{\mathrm{cc}}$ )
3. Note down the output voltage at CRO
4. To observe the phase difference between the input and the output, set the CRO in dual Mode and switch the trigger source in CRO to CHI.
5. Plot the input and output waveforms on the graph.

## Result:

Thus Astable \& Monostable Multivibrators and schmitt trigger were designed using opamp and the waveforms were plotted.

## 6. RC PHASE SHIFT AND WIEN BRIDGE OSCILLATORS USING OP-AMP.

Aim:
To design the following sine wave oscillators
a) Wein Bridge Oscillator with the frequency of 1 KHz .
b) RC Phase shift oscillator with the frequency of 200 Hz .

## Components Required:

| S.No | Components | Range | Quantity |
| :---: | :--- | :--- | :---: |
| 1. | Op-amp | IC 741 | 1 |
| 2. | Dual trace supply | $(0-30) \mathrm{V}$ | 1 |
| 3. | Function Generator | $(0-2) \mathrm{MHz}$ | 1 |
| 4. | Resistors |  |  |
| 5. | Capacitors |  |  |
| 6 | CRO | $(0-30) \mathrm{MHz}$ | 1 |
| 7 | Probes | -- | -- |

## Equations Related to the Experiments:

a) Wein Bridge Oscillator

Closed loop gain $\mathrm{A}_{\mathrm{v}}=\left(1+\mathrm{R}_{\mathrm{f}} / \mathrm{R}_{1}\right)=3$
Frequency of Oscillation $f_{a}=1 /(2 \pi R C)$
b) RC Phase shift Oscillator:

Gain $\mathrm{A}_{\mathrm{v}}=\left[\mathrm{R}_{\mathrm{f}} / \mathrm{R}_{1}\right]=29$
Frequency of oscillation $\mathrm{f}_{\mathrm{a}}=1 \sqrt{6} * 2 * \pi * \mathrm{RC}$

1) Wein Bridge Oscillator:

## Design:

Gain required for sustained oscillation is $\mathrm{A}_{\mathrm{v}}=1 / \beta=3$
(PASS BAND GAIN) (i.e.) $1+\mathrm{R}_{\mathrm{f}} / \mathrm{R}_{1}=3$
$\therefore \mathrm{R}_{\mathrm{f}}=2 \mathrm{R}_{1}$
Frequency of Oscillation $f_{0}=1 / 2 \pi$ R C
Given $\mathrm{f}_{\mathrm{o}}=1 \mathrm{KHz}$
Let $\mathrm{C}=0.05 \mu \mathrm{~F}$
$\therefore \mathrm{R}=1 / 2 \pi \mathrm{f}_{\mathrm{o}} \mathrm{C}$

$$
\mathrm{R}=3.2 \mathrm{~K} \Omega
$$

Let $\mathrm{R} 1=10 \mathrm{~K} \Omega \quad \therefore \mathrm{Rf}=2 * 10 \mathrm{~K} \Omega$

Circuit diagram:


Tabulation:

| Derived frequency <br> $(\mathrm{Hz})$ | Observed frequency <br> $(\mathrm{Hz})$ | Amplitude (V) | Time period (ms) |
| :---: | :---: | :---: | :---: |
|  |  |  |  |

## Model Graph:



## Procedure:

1. Connect the components as shown in the circuit
2. Switch on the power supply and CRO.
3. Note down the output voltage at CRO.
4. Plot the output waveform on the graph.
5. Redesign the circuit to generate the sine wave of frequency 2 KHz .
6. Compare the output with the theoretical value of oscillation.

## Observation:

Peak to peak amplitude of the output $=\quad$ Volts.
Frequency of oscillation $=\mathrm{Hz}$.

## 2) RC Phase Shift Oscillators:

## Design:

Frequency of oscillation fo $=1 /\left(\sqrt{6} * 2 * \Pi^{*} \mathrm{RC}\right)$
$\mathrm{Av}=[\mathrm{Rf} / \mathrm{R} 1]=29$
$\mathrm{R}_{1}=10 \mathrm{R}$
$\mathrm{R}_{\mathrm{f}}=29 \mathrm{R}_{1}$
Given fo $=200 \mathrm{~Hz}$.

$$
\text { Let } \mathrm{C}=0.1 \mu \mathrm{~F}
$$

$$
\begin{aligned}
\mathrm{R} & =1 /\left(\sqrt{6} * 2 \pi * \mathrm{fo}^{*} \mathrm{C}\right) \\
& =1 /\left(\sqrt{6} * 2 * \pi * 200 * 0.1 * 10^{-6}\right) \\
& =\quad \mathrm{K} \Omega
\end{aligned}
$$

To prevent the loading of amplifier by RC network, $\mathrm{R} 1 \geq 10 \mathrm{R}$

$$
\therefore \mathrm{R} 1=10^{*}------\quad=\mathrm{K} \Omega
$$

Since $R f=29 R 1$

$$
\begin{gathered}
\mathrm{Rf}=29^{*}------ \\
=\quad \mathrm{M} \Omega
\end{gathered}
$$

## Circuit diagram:



## Tabulation:

| Derived frequency <br> $(\mathrm{Hz})$ | Observed frequency <br> $(\mathrm{Hz})$ | Amplitude (V) | Time period (ms) |
| :---: | :---: | :---: | :---: |
|  |  |  |  |

## Model Graph:



## Procedure:

1. Connect the circuits as shown in the circuit 5.2
2. Switch on the power supply Note down the output voltage on the CRO.
3. Plot the output waveforms on the graph.
4. Redesign the circuit to generate the sine wave of 1 KHz .
5. Plot the output waveform on the graph.
6. Compare the practical value of the frequency with the theoretical value.

## Observation:

$$
\begin{aligned}
& \text { Peak to peak amplitude of the sine wave }= \\
& \text { Frequency of Oscillation (obtained) } \quad=
\end{aligned} \quad \text { Volts } . \mathrm{Hz} . ~ l i
$$

## Result:

Thus wien bridge oscillator and RC Phase shift oscillator was designed using op-amp and tested.

## Aim:

To design and test an Astable and Monostable Multivibrators using 555 timer with duty cycles ratio.

## Apparatus Required:

| S.No | Component | Range | Quantity |
| :---: | :--- | :--- | :---: |
| 1. | 555 TIMER |  | 1 |
| 2. | Resistors | $3.3 \mathrm{~K}, 6.8 \mathrm{k}$ | 1 |
| 3. | Capacitors | $0.1 \mu \mathrm{~F}, 0.01 \mu \mathrm{~F}$ | 2 |
| 4. | Diode | In 4001 | 1 |
| 5. | CRO |  | 1 |
| 6. | Power supply | $\pm 15 \mathrm{~V}$ | 1 |
| 7. | Probe |  | 2 |
| 8. | Bread Board |  | 1 |

## Astable Multivibrators using 555

Fig shows the 555 timer connected as an Astable Multivibrators. Initially, when the output is high. Capacitor C starts charging towards $\mathrm{V}_{\mathrm{cc}}$ through $\mathrm{R}_{\mathrm{A}}$ and $\mathrm{R}_{\mathrm{B}}$. As soon as capacitor voltage equals $2 / 3 \mathrm{~V}_{\mathrm{cc}}$ upper comparator (UC) triggers the flip flop and the output switches low. Now capacitor C starts discharging through $\mathrm{R}_{\mathrm{B}}$ and transistor $\mathrm{Q}_{1}$.

When the voltage across $C$ equals $1 / 3 \mathrm{~V}_{\mathrm{cc}}$ lower comparator (LC), output triggers the flipflop and the output goes high. Then the cycle repeats.

The capacitor is periodically charged and discharged between $2 / 3 \mathrm{~V}_{\mathrm{cc}}$ and $1 / 3 \mathrm{~V}_{\mathrm{cc}}$ respectively. The time during which the capacitor charges form $1 / 3 \mathrm{~V}_{\mathrm{cc}}$ to $2 / 3 \mathrm{~V}_{\mathrm{cc}}$ is equal to the time the output is high and is given by

$$
\begin{equation*}
\mathrm{T}_{\mathrm{c}}=0.69\left(\mathrm{R}_{\mathrm{A}}+\mathrm{R}_{\mathrm{B}}\right) \mathrm{C} \tag{1}
\end{equation*}
$$

Where $\mathrm{R}_{\mathrm{A}}$ and $\mathrm{R}_{\mathrm{B}}$ are in Ohms and C is in farads. Similarly the time during which the capacitor discharges from $2 / 3 \mathrm{~V}_{\mathrm{cc}}$ to $1 / 3 \mathrm{~V}_{\mathrm{cc}}$ is equal to the time the output is low and is given by

$$
\begin{equation*}
\mathrm{T}_{\mathrm{d}}=0.69 \mathrm{R}_{\mathrm{B}} \mathrm{C} \tag{2}
\end{equation*}
$$

The total period of the output waveform is

$$
\begin{equation*}
T=T_{c}+T_{d}=0.69\left(R_{A}+2 R_{B}\right) C \tag{3}
\end{equation*}
$$

The frequency of oscillation

$$
\begin{equation*}
\mathrm{f}_{\mathrm{o}}=1 / \mathrm{T}=1.45 /\left(\mathrm{R}_{\mathrm{A}}+2 \mathrm{R}_{\mathrm{B}}\right) \mathrm{C} \tag{4}
\end{equation*}
$$

Eqn (4) shows that fo is independent of supply voltage Vcc
The duty cycle is the ratio of the time $\mathrm{t}_{\mathrm{d}}$ during which the output is low to the total time period T. This definition is applicable to 555 Astable Multivibrators only; conventionally the duty cycle ratio is defined as the ratio as the time during which the output is high to the total time period.
$\therefore$ Duty cycle $=\mathrm{t}_{\mathrm{d}} \mathrm{T} \times 100$

$$
\begin{equation*}
\mathrm{R}_{\mathrm{B}}+\mathrm{R}_{\mathrm{A}}+2 \mathrm{R}_{\mathrm{B}} \times 100 \tag{5}
\end{equation*}
$$

To obtain $50 \%$ duty cycle a diode should be connected across $\mathrm{R}_{\mathrm{B}}$ and $\mathrm{R}_{\mathrm{A}}$ must be a combination of a fixed resistor and a potentiometer. So that the potentiometer can be adjusted for the exact square waves

## DESIGN:

Design an Astable Multivibrators for a frequency of $\qquad$ KHz with a duty cycle ratio of D $=50 \%$

$$
f 0=1 / T \quad=1.45 /\left(R_{A}+2 R_{B}\right) C
$$

Choosing $\mathrm{C}=1 \mu \mathrm{~F} ; \mathrm{R}_{\mathrm{A}}=560$

$$
\begin{gathered}
\mathrm{D}=\mathrm{R}_{\mathrm{B}} / \mathrm{R}_{\mathrm{A}}+2 \mathrm{R}_{\mathrm{B}}=0.5[50 \%] \\
\mathrm{R}_{\mathrm{B}}=
\end{gathered}
$$

## Circuit Diagram



Tabulation:

## Model Graph



## Procedure:

1. Rig-up the circuit of 555 Astable Multivibrators as shown in fig with the designed value of components.
2. Connect the CRO probes to pin 3 and 2 to display the output signal and the voltage across the timing capacitor. Set suitable voltage sensitively and time-base on the CRO.
3. Switch on the power supply to CRO and the circuit.
4. Observe the waveforms on the CRO and draw to scale on a graph sheet. Measure the voltage levels at which the capacitor starts charging and discharging, output high and low timings and frequency.
5. Switch off the power supply. Connect a diode across $R_{B}$ as shown in dashed lines in fig to make the Astable with $50 \%$ duty cycle ratio. Switch on the power supply. Observe the output waveform. Draw to scale on a graph sheet.

## Monostable Multivibrators using 555

Monostable Multivibrators has one stable state and other is a quasi stable state. The circuit is useful for generating single output pulse at adjustable time duration in response to a triggering signal. The width of the output pulse depends only on external components, resistor and a capacitor.

The stable state is the output low and quasi stable state is the output high. In the stable state transistor Q1 is 'on' and capacitor C is shorted out to ground. However upon application of
a negative trigger pulse to pin2, Q1 is turned 'off' which releases the short circuit across the external capacitor C and drives the output high. The capacitor C now starts charging up towards $\mathrm{V}_{\mathrm{cc}}$ through $\mathrm{R}_{\mathrm{A}}$. However when the voltage across C equal $2 / 3 \mathrm{~V}_{\mathrm{cc}}$ the upper comparator output switches form low to high which in turn drives the output to its low state via the output of the flip flop. At the same time the output of the flip flop turns Q1 'on' and hence C rapidly discharges through the transistor. The output remains low until a trigger is again applied. Then the cycle repeats.

The pulse width of the trigger input must be smaller than the expected pulse width of the output. The trigger pulse must be of negative going signal with amplitude larger than $1 / 3$ Vcc. The width of the output pulse is given by,

$$
\mathrm{T}=1.1 \mathrm{R}_{\mathrm{A}} \mathrm{C}
$$

## Design:

Given a pulse width of duration of $100 \mu \mathrm{~s}$
Let $\mathrm{C}=0.01 \mathrm{mfd} ; \mathrm{F}=$ $\qquad$ KHz

Here, $\mathrm{T}=1.1 \mathrm{R}_{\mathrm{A}} \mathrm{C}$
So, $\mathrm{R}_{\mathrm{A}}=$

## Circuit Diagram:



## Model Diagram:

## Procedure:

1. Rig-up the circuit of 555 monostable Multivibrators as shown in fig with the designed value of components.
2. Connect the trigger input to pin 2 of 555 timer form the function generator.
3. Connect the CRO probes to pin 3 and 2 to display the output signal and the voltage across the timing capacitor. Set suitable voltage sensitively and time-base on the CRO.
4. Switch on the power supply to CRO and the circuit.
5. Observe the waveforms on the CRO and draw to scale on a graph sheet. Measure the voltage levels at which the capacitor starts charging and discharging, output high and low timings along with trigger pulse.

## Observation:

Peak to peak amplitude of the output $=\quad$ Volts.
Frequency $=\quad \mathrm{Hz}$.
Upper threshold voltage $=\quad$ Volts.
Lower threshold voltage $=$ Volts.

## Result:

Thus the Astable Multivibrators and Monostable Multivibrators using 555 timer is designed and tested.

## 9. DC POWER SUPPLY USING LM317 AND LM723.

## AIM :

To design a high current, low voltage and high voltage linear variable dc regulated power supply and test its line and load regulation.

## COMPONENTS REQUIRED :

| S.NO | COMPONENTS | SPECIFICATION | QUANTITY |
| :--- | :--- | :--- | :--- |
| 1. | Transistors | TIP122,2N3055 | 1 each |
| 2. | Integrated Circuit | LM723 | 1 |
| 3. | Digital Ammeter | $(0-10) \mathrm{A}$ | 1 |
| 4. | Digital Voltmeter | $(0-20) \mathrm{V}$ | 1 |
| 5. | Variable Power Supply | $(0-30) \mathrm{V}-2 \mathrm{~A}$ | 1 |
| 6. | Resistors | $300 \Omega, 430 \Omega, 1 \mathrm{~K} \Omega, 678 \mathrm{~K} \Omega, 678 \Omega$ <br> $1 \Omega$ | 1 each <br> 2 |
| 7. | Capacitors | $0.1 \mu \mathrm{~F}, 100 \mathrm{pF}$ | 1 each |
| 9. | Rheostat | $(0-350) \Omega$ | 1 |

## CIRCUIT DIAGRAM: Low Voltage Regulator



## DESIGN:

Output voltage $\rightarrow \mathrm{V}_{\mathrm{O}}$
Reference voltage $\rightarrow$ Vref
Rprotect $\quad \rightarrow$ Minimum Resistance to protect the output from short circuit.

## Low Voltage Regulator :

Given : Vo=5V, Vref $=\mathbf{7 . 1 5} \mathrm{V}$
To calculate R1, R2,R3 and Rsc.
$\operatorname{Vo}=\operatorname{Vref}(\mathrm{R} 2 /(\mathrm{R} 1+\mathrm{R} 2))$
$5 / 7.15=(\mathrm{R} 2 /(\mathrm{R} 1+\mathrm{R} 2))$
( R1 + R2 ) $0.699=\mathrm{R} 2$
$0.699 \mathrm{R} 1=0.301 \mathrm{R} 2, \mathrm{R} 1=0.4306 \mathrm{R} 2$
Select $\mathbf{R 2}=\mathbf{1} \mathbf{K} \boldsymbol{\Omega}$
$\mathrm{R} 1=1 \mathrm{~K} \Omega * 0.4306=430 \Omega$
$R 1=430 \Omega$

$$
\begin{aligned}
& \mathrm{R} 3=\mathrm{R} 1 * \mathrm{R} 2 /(\mathrm{R} 1+\mathrm{R} 2), \mathrm{R} 3=430.6 * 1000 /(430.6+1000) \\
& \mathbf{R} 3=\mathbf{3 0 0 \Omega} \\
& \mathrm{Rsc}=\mathrm{V}_{\text {sense }} / \mathrm{I}_{\text {limit }}=0.5 / 1 \mathrm{~A}=0.5 \Omega, \mathbf{R s c}=\mathbf{0 . 5 \Omega}
\end{aligned}
$$

## High Voltage Regulator :

| Sl. No. | Name of item \& specifications | Quantity |
| :---: | :--- | :--- |
| 1 | Multimeter | 2 |
| 2 | Resistor | 1 |
| 3 | Potentiometer | 1 |
| 4 | Capacitor | 3 |
| 4 | IC LM 317 | 1 |
| 5 | Transformer (230 / 12 V) | 1 |
| 7. | Diode 1N4001 | 4 |
| 8 | Bread board | 1 |
| 9 | Connecting Wires | As required |

## CIRCUIT DIAGRAM:

High Voltage Regulator:


REGULATOR PINOUT


## High Voltage Regulator :

## Given : Vo $=12 \mathrm{~V}$, Vref $=7.15 \mathrm{~V}$

To calculate R1, R2,R3 and Rsc.
$\operatorname{Vo}=\operatorname{Vref}(1+(\mathrm{R} 1 / \mathrm{R} 2))$
$12 / 7.15=1+(\mathrm{R} 1 / \mathrm{R} 2)$
$(12 / 7.15)-1=(\mathrm{R} 1 / \mathrm{R} 2)$
$(\mathrm{R} 1 / \mathrm{R} 2)=0.678$
Select $\mathbf{R 2}=\mathbf{1} \mathbf{K} \boldsymbol{\Omega}$
$\mathrm{R} 1=1 \mathrm{~K} \Omega * 0.678=678 \Omega$
$R 1=678 \Omega$

Rsc $=V_{\text {sense }} / I_{\text {limit }}=0.5 / 1 \mathrm{~A}=0.5 \Omega$
$\mathbf{R s c}=\mathbf{0 . 5 \Omega}$

Tabulation of the Measurements :

## LOW VOLTAGE REGULATOR :

Line Regulation :
Load Regulation

| S.No. | Load Resistance $\mathrm{R}_{\mathrm{L} 1}=$ |  |  | Input Voltage $\mathrm{V}_{\text {in } 1}=$ |  |
| :--- | :--- | :--- | :--- | :--- | :---: |
|  | Input Voltage <br> Vin(Volts) | Output <br> Voltage <br> $\mathrm{V}_{\mathrm{L}}$ (Volts) | Input Voltage <br> Vin(Volts) | Output Voltage <br> $\mathrm{V}_{\mathrm{L}}$ (Volts) |  |
|  |  |  |  |  |  |
|  |  |  |  |  |  |

## HIGH VOLTAGE REGULATOR :

## Line Regulation :

Load Regulation :

| S.No. | Load Resistance $\mathrm{R}_{\mathrm{L} 1}=$ |  | Input Voltage $\mathrm{V}_{\text {in } 1}=$ |  |
| :--- | :--- | :--- | :--- | :--- |
|  | Input Voltage <br> Vin(Volts) | Output Voltage <br> $\mathrm{V}_{\mathrm{L}}$ (Volts) | Input Voltage <br> $\mathrm{I}_{\mathrm{L}}(\mathrm{A})$ | Output Voltage <br> $\mathrm{V}_{\mathrm{L}}$ (Volts) |
|  |  |  |  |  |
|  |  |  |  |  |

## Calculation of \% Voltage Regulation :

\% Voltage Regulation $=\left(\mathrm{V}_{\mathrm{dc}}(\mathrm{NL})-\mathrm{V}_{\mathrm{dc}}(\mathrm{FL})\right) / \mathrm{V}_{\mathrm{dc}}(\mathrm{FL})$
$\mathrm{V}_{\mathrm{dc}}(\mathrm{NL})=$ D.C. output voltage on no load
$\mathrm{V}_{\mathrm{dc}}(\mathrm{FL})=$ D.C. output voltage on full load

## Model Graph :

Line Regulation :
Input Voltage Vs Output Voltage :



## PROCEDURE :

## LOW VOLTAGE REGULATOR :

## Line Regulation :

1. Give the circuit connection as per the circuit diagram shown in Fig 1.1.
2. Set the load Resistance to give load current of 0.25 A .
3. Vary the input voltage from 7 V to 18 V and note down the corresponding output voltages.
4. Similarly set the load current ( $\mathrm{I}_{\mathrm{L}}$ ) to $0.5 \mathrm{~A} \& 0.9 \mathrm{~A}$ and make two more sets of measurements.

## Load Regulation :

1. Set the input voltage to 10 V .
2. Vary the load resistance in equal steps from $350 \Omega$ to $5 \Omega$ and note down the corresponding output voltage and load current.
3.Similarly set the input voltage ( Vin ) to 14 V \& 18 V and make two more sets of measurements.

## Lab Report :

1.Plot the line regulation by taking Input Voltage (Vin) along X-axis and Output Voltage $\left(\mathrm{V}_{\mathrm{L}}\right)$ along Y -axis for various load currents.
2.Plot the load regulation by taking load current $\left(\mathrm{I}_{\mathrm{L}}\right)$ along X -axis and Output Voltage $\left(\mathrm{V}_{\mathrm{L}}\right)$ along Y -axis for various input voltages.
3.Calculate its \% Voltage Regulation using the formula.

## HIGH VOLTAGE REGULATOR :

## Line Regulation :

1.Give the circuit connection as per the circuit diagram shown in Fig 1.2.
2.Set the load Resistance to give load current $\mathrm{I}_{\mathrm{L}}$ of 0.25 A .
3.Vary the input voltage from 7 V to 18 V and note down the corresponding output voltages.
4.Similarly set the load current ( $\mathrm{I}_{\mathrm{L}}$ ) to $0.5 \mathrm{~A} \& 0.9 \mathrm{~A}$ and make two more sets of measurements.

## Load Regulation :

1. Set the input voltage to 10 V .
2. Vary the load resistance in equal steps from $350 \Omega$ to $15 \Omega$ and note down the corresponding output voltage and load current.
3.Similarly set the input voltage ( Vin ) to $14 \mathrm{~V} \& 18 \mathrm{~V}$ and make two more sets of measurements.

## Lab Report :

1.Plot the line regulation by taking Input Voltage (Vin) along X-axis and Output Voltage $\left(\mathrm{V}_{\mathrm{L}}\right)$ along Y-axis for various load currents.
2.Plot the load regulation by taking load current $\left(\mathrm{I}_{\mathrm{L}}\right)$ along X -axis and Output Voltage $\left(\mathrm{V}_{\mathrm{L}}\right)$ along Y-axis for various input voltages.
3.Calculate its \% Voltage Regulation using the formula.

## Result :

Thus the line and load regulation of a high current, low voltage and high voltage linear variable dc regulated power supply was designed and tested.

## 10-BIT R-2R DIGITAL TO ANALOG CONVERTER

Aim : To design 4 bit R-2R ladder DAC using Op-Amp for an output voltage of 5 V when the input is 10 (Binary 1010).

## Apparatus :

| S. <br> No. | Particulars | Specification | Quantity |
| :---: | :--- | :---: | :---: |
| 1. | IC | $\mu \mathrm{A} 741$ | 02 |
| 2. | Resistors | As per design | - |
| 3. | Multimeter | - | 01 |
| 4. | Base board + connecting wires | - | 01 Set |

## Procedure :

1. Connections are made as shown in the circuit diagram.
2. Digital input data is given at $\mathrm{D} 3, \mathrm{D} 2, \mathrm{D} 1, \mathrm{D} 0$ and corresponding analog output voltage $\mathrm{V}_{0}$ is measured.
3. Tabulate the readings \& plot the graph between $V_{o}$ on $y$-axis $V_{\text {in }}$ on $X$-axis.

## Note :

1. D0.D1.D2 \& D3 are binary input.
2. Vo is the analog output.
3. Binary inputs Do.D1.D2 \& D3 can take either the value ' 0 ' or ' 1 '.
4. Binary input $\operatorname{Di}(i=0$ to 3 ) can be made ' 0 ' by connecting the $\mathrm{i} / \mathrm{p}$ to ground. It can be made ' 1 ' by connecting to -5 V .

Logic $0 \rightarrow 0 \mathrm{~V}$
Logic $1 \rightarrow+5 \mathrm{~V}$

## Circuit Diagram



Result :

| $\begin{aligned} & \text { Decimal } \\ & \text { Value } \end{aligned}$ | Binary Inputs |  |  |  | Analog O/P Vo(volts) <br> Theoretical values | Analog O/P Vo(volts) <br> Practical values |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | D3 | $\mathrm{D}_{2}$ | D1 | D0 |  |  |
| 0 | 0 | 0 | 0 | 0 |  |  |
| 1 | 0 | 0 | 0 | 1 |  |  |
| 2 | 0 | 0 | 1 | 0 |  |  |
| 3 | 0 | 0 | 1 | 1 |  |  |
| 4 | 0 | 1 | 0 | 0 |  |  |
| 5 | 0 | 1 | 0 | 1 |  |  |
| 6 | 0 | 1 | 1 | 0 |  |  |
| 7 | 0 | 1 | 1 | 1 |  |  |
| 8 | 1 | 0 | 0 | 0 |  |  |
| 9 | 1 | 0 | 0 | 1 |  |  |
| 10 | 1 | 0 | 1 | 0 |  |  |
| 11 | 1 | 0 | 1 | 1 |  |  |
| 12 | 1 | 1 | 0 | 0 |  |  |
| 13 | 1 | 1 | 0 | 1 |  |  |
| 14 | 1 | 1 | 1 | 0 |  |  |
| 15 | 1 | 1 | 1 | 1 |  |  |

## TWO BIT FLASH ADC

Aim : To realize two bit Flash ADC
Apparatus :

| Sl. | Particulars | Specification | Quantity |
| :---: | :---: | :---: | :---: |
| No. |  | LM324 <br> 7404 <br> 7432 <br> 7409 |  |
|  | IC | 1 K | 04 |
| 2. | Resistors | - | 04 |
| 3. | Multimeter | - | 01 |
| 4. | Base board + connecting wires | - | 01 Set |

## Circuit diagram



## Procedure :

1. Connections are made as shown in the circuit diagram.
2. Apply the different analog input voltages and note down the output.

## Tabulation and Result

| Analog Input | A | B | C Digital Output |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  |  |  |  |
| 0 to $\frac{1}{4} V c c$ | 0 | 0 | 0 | 0 | 0 |
| $\frac{1}{4} V c c$ to $\frac{2}{4} V c c$ | 0 | 0 | 1 | 0 | .$c 9$ |
| $\frac{2}{4} V c c$ to $\frac{3}{4} V c c$ | 0 | 0 | 1 | 1 | 1 |
| $\frac{3}{4} V c c$ to $V c c$ | 1 | 1 | 1 | 1 | 1 |

## 10. STUDY OF SMPS.

## Aim:

To study the switched-mode power supply.

## Apparatus required:

switched-mode power supply.

## Block Diagram:

A switched-mode power supply (switching-mode power supply/SMPS, or simply switcher) is an electronic power supply unit (PSU) that incorporates a switching regulator in order to provide the required output voltage. An SMPS is a power converter that transmits power from a source (e.g., a battery or the electrical power grid) to a load (e.g., a personal computer). The function of the converter is to provide a regulated output voltage usually at a different level from the input voltage.

## Block Diagram:



Block diagram of a mains operated AC-DC SMPS with output voltage regulation

## Input rectifier stage



AC, half-wave and full-wave rectified signals.
If the SMPS has an AC input, then the first stage is to convert the input to DC. This is called rectification. The rectifier circuit can be configured as a voltage doubler by the addition of a switch operated either manually or automatically. This is a feature of larger supplies to permit operation from nominally 120 V or 240 V supplies. The rectifier produces an unregulated DC voltage which is then sent to a large filter capacitor. The current drawn from the mains supply by this rectifier circuit occurs in short pulses around the AC voltage peaks.

## Inverter stage

The inverter stage converts DC, whether directly from the input or from the rectifier stage described above, to AC by running it through a power oscillator, whose output transformer is very small with few windings at a frequency of tens or hundreds of kilohertz ( kHz ). The frequency is usually chosen to be above 20 kHz , to make it inaudible to humans. The output voltage is optically coupled to the input and thus very tightly controlled.

## Voltage converter and output rectifier

If the output is required to be isolated from the input, as is usually the case in mains power supplies, the inverted AC is used to drive the primary winding of a high-frequency transformer. This converts the voltage up or down to the required output level on its secondary winding. The output transformer in the block diagram serves this purpose.

## Result:

Thus the switched-mode power supply was studied.

## 11. SIMULATION OF EXPERIMENTS 3, 4, 5, 6 AND 7 USING PSPICE NETLISTS.


#### Abstract

Aim: To simulate the following experiments using Pspice Netlist software. 1. Instrumentation amplifier. 2. Active Lowpass, Highpass and Bandpass filters. 3. Astable and Monostable multivibrators and Schmitt trigger using Op-Amp. 4. Phase shift and Wien bridge oscillators using Op-Amp.

\section*{Apparatus Required:}

Personal computer workstation available in the laboratory. PSpice circuit simulator and Probe graphic post processor software available on the PC.


## Theory:

PSPICE (Simulation Program with Integrated Circuit Emphasis).is a general purpose circuit simulation programming language that performs non-linear DC, transient and linear steady state AC analysis. Linear circuit elements of resistance, capacitance, inductance, independent and dependent current and voltage sources are modeled.

Personal

- To start PSpice, double click on the PSpice schematic icon or select schematic from the design eval software menu:
- Draw schematics:

1. Select the desired component from DRAW/GET NEW PART and place the part.
2. Repeat the first step until all necessary parts are placed.
3. Connect each component with the DRAW/WIRE command.
4. Save the Circuit with your initials in your directory.
5. Select desired analysis and set the options (ANALYSIS/SETUP)
6. Run PSpice (ANALYSIS/RUN PSPICE)

- If errors are found, they will be described in the $m y_{-}$file. OUT file after execution. View the contents of this file using ANALYSIS/EXAMINE OUTPUT.
- During this experiment, hard copy output will not be required.


## PROCEDURE:

1. For each circuit perform the following:

- Draw the circuit scematic in Pspice, label all the nodes.
- Setup the analysis for a DC sweep from -5 V to 5 V in 0.1 V increments.
- Run the simulation. Display the correct I-V graph in Probe.

4. Record the annotated circuit diagram, the simulation results, graphs and your comments in your laboratory notebook. Have your instructor inspect

## Result:

The experiments were simulated using Pspice and the results were Compared.

